

Today

→ OFDM

→ Motivation

→ Packet Detection

→ Cyclic Prefix

→ CFO

→ Channel Estimation

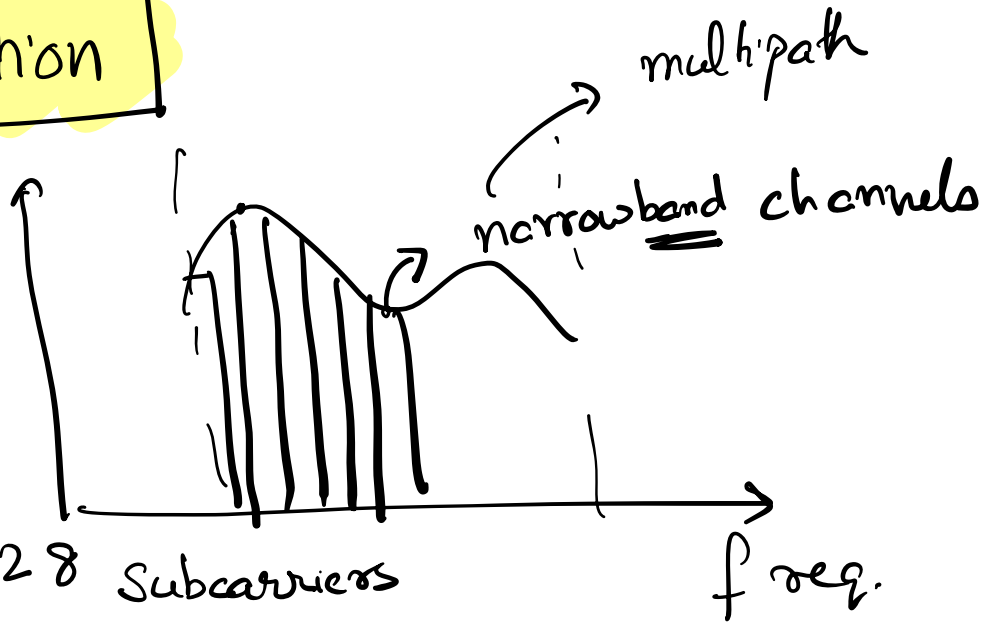
→ Residual CFO & SFO

# Motivation

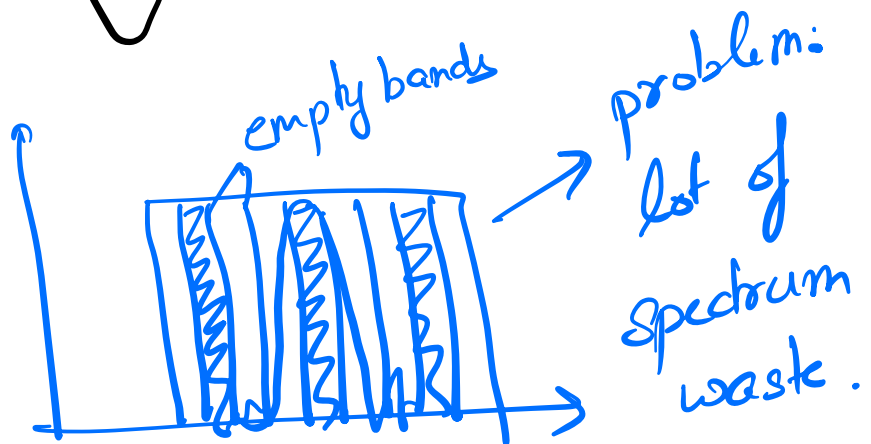
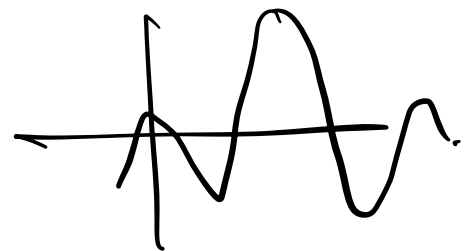
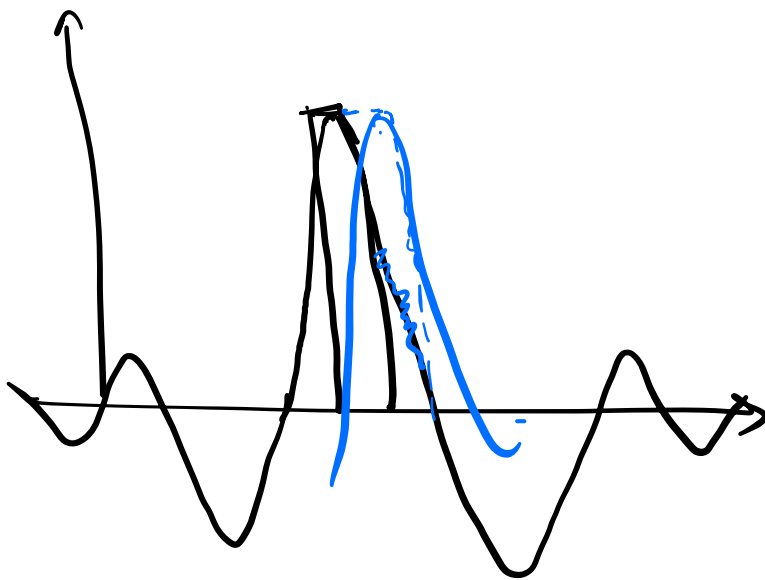
Wi-Fi

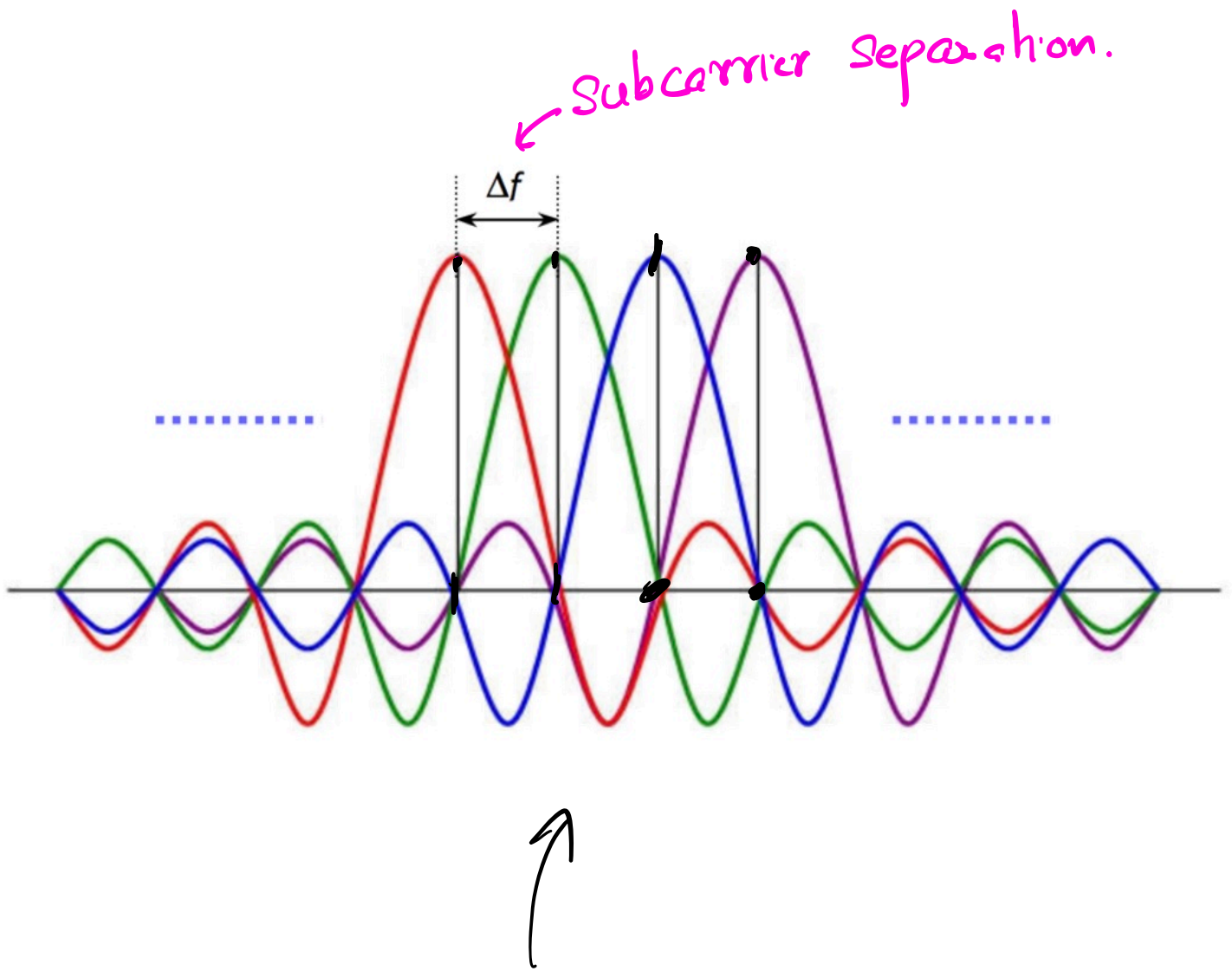
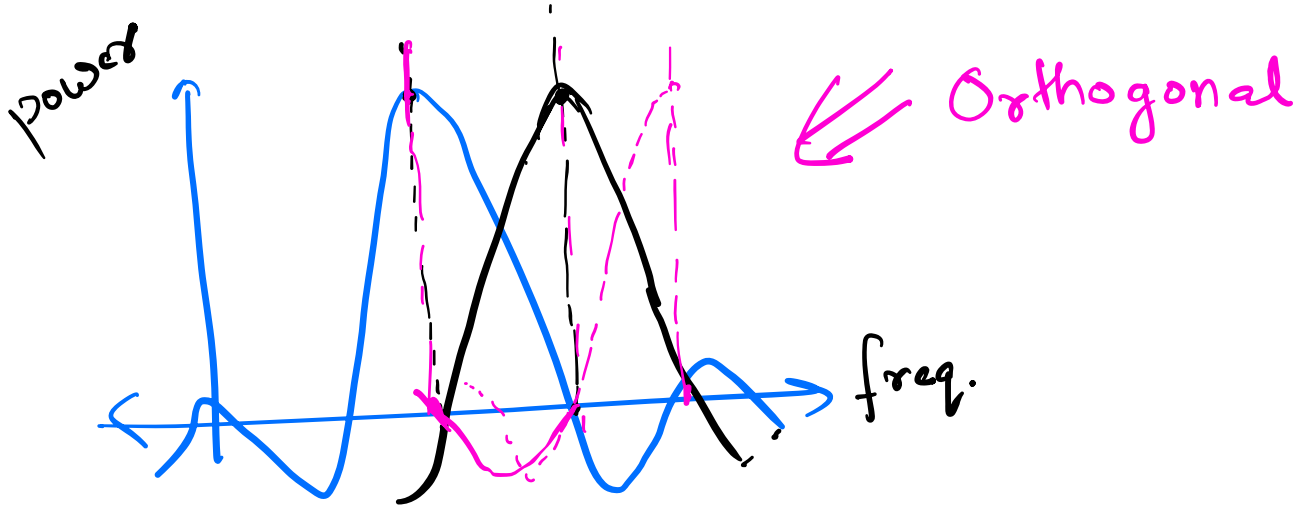
40 MHz  $\rightarrow$  128 subcarriers

20 MHz  $\rightarrow$  64 subcarriers



OFDM: Orthogonal Freq. Division Multiplexing.





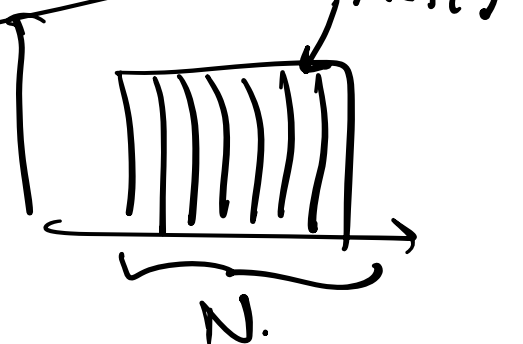
# Fourier Transform

time domain  $\rightleftharpoons$  frequency domain

## Discrete Fourier Transform

$x(t) \rightarrow X(f)$  [N samples]  $\leftarrow$  DFT  
 $\rightarrow X(f_i) = \frac{1}{N} \sum_{t=0}^{N-1} x(t) e^{-j2\pi f_i t / N}$

$\underline{x(t)} = \sum_{i=0}^{N-1} X(f_i) e^{j2\pi f_i t / N}$   $\leftarrow$  IDFT

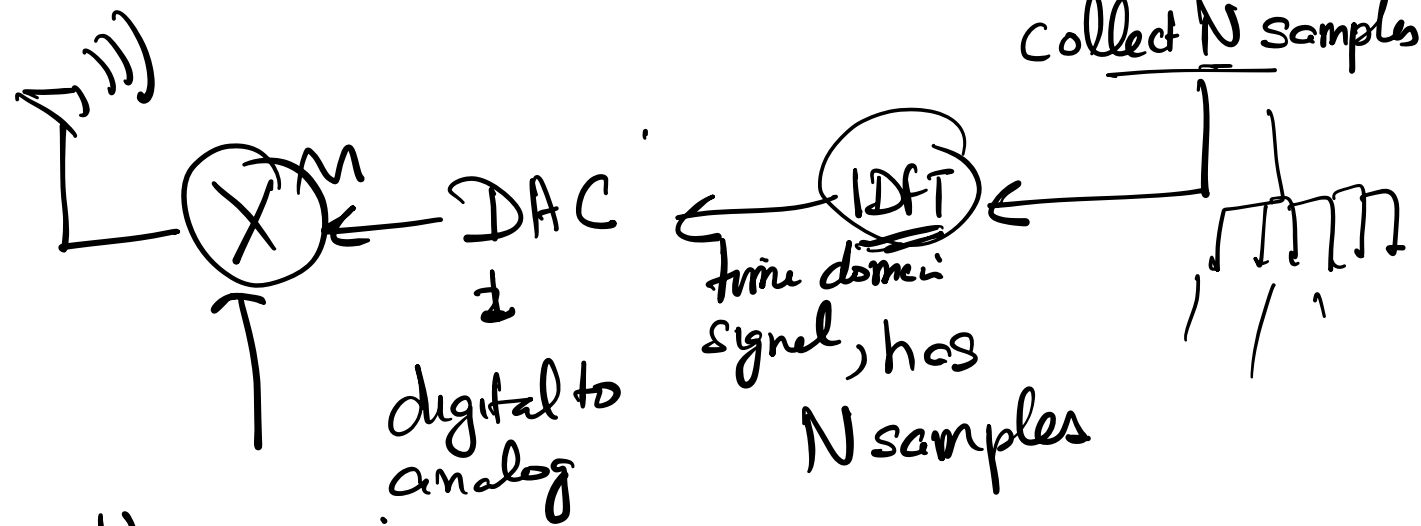
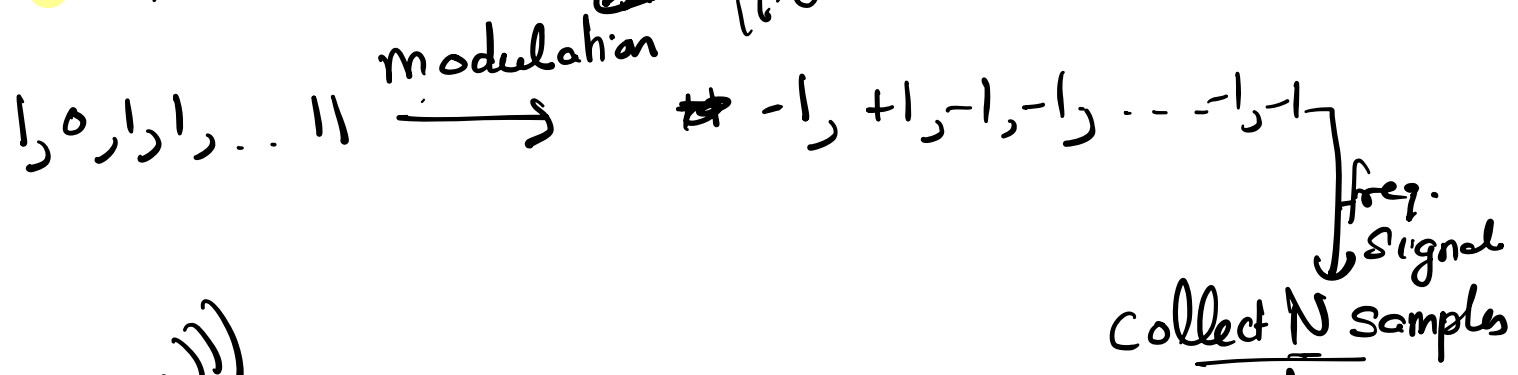


$1, 0, 1, \dots, 1, 1 \rightarrow$  collect N samples  $\rightarrow$  IDFT  
 $\downarrow$   
 time domain signal  $x(t)$

$\dots, 1, 1, 0, 1, 1 \leftarrow$  DFT  $\leftarrow$  collect N samples

# Pipeline

Transmitter.



"complex math"

$\rightarrow$  Bandwidth (20MHz),  $N$  samples - IDFT  $\xrightarrow{1024}$  1ms

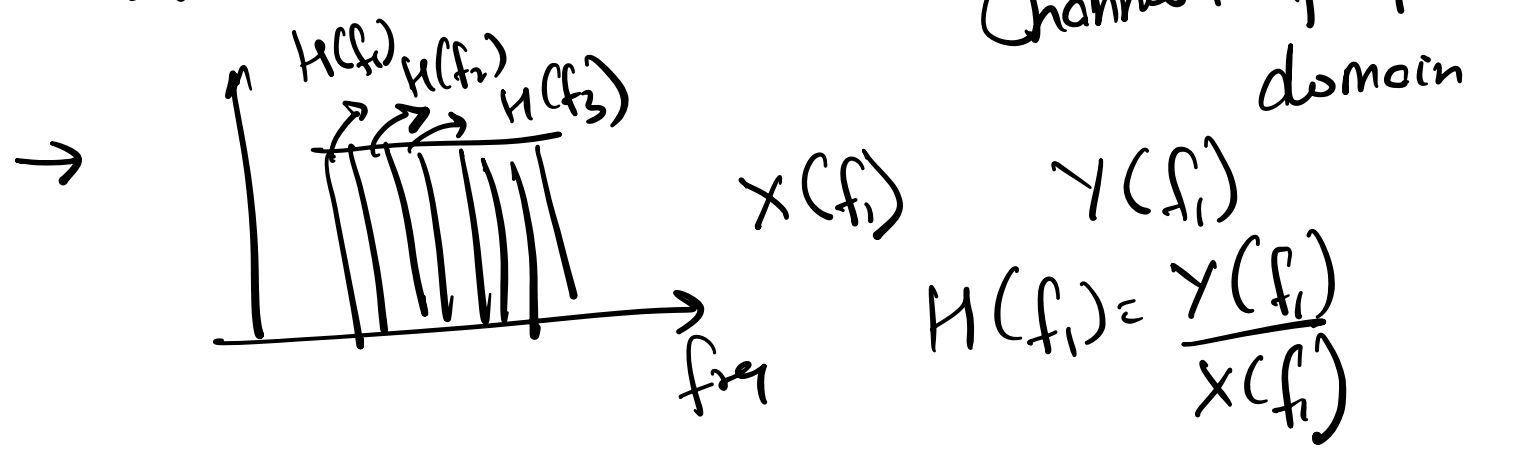
$\frac{20\text{MHz}}{64} = \frac{20\text{MHz}}{N} \rightarrow 64$  samples

$\frac{1\text{Mbps}}{1024} \rightarrow 1\text{ms}$

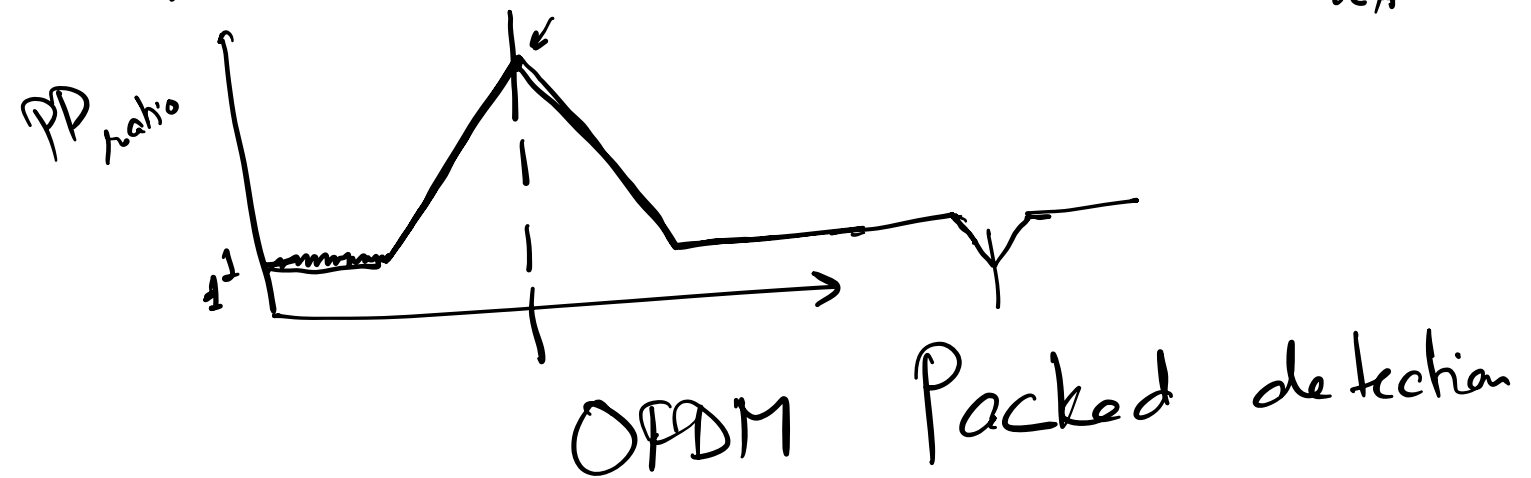
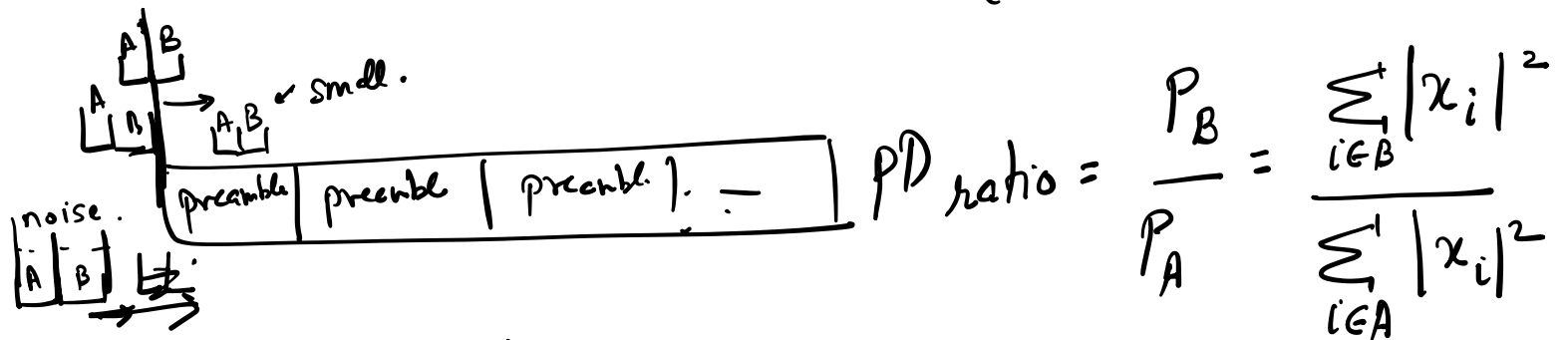
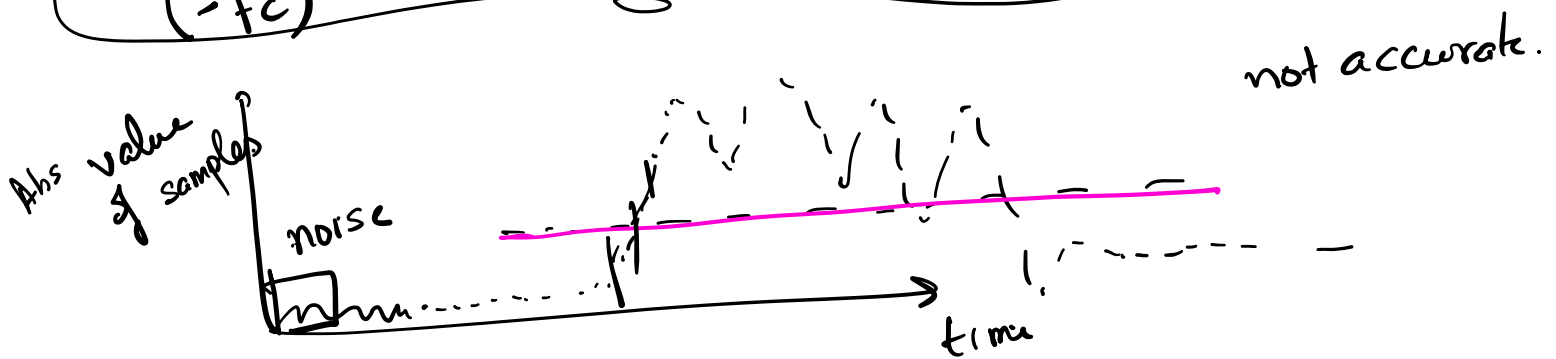
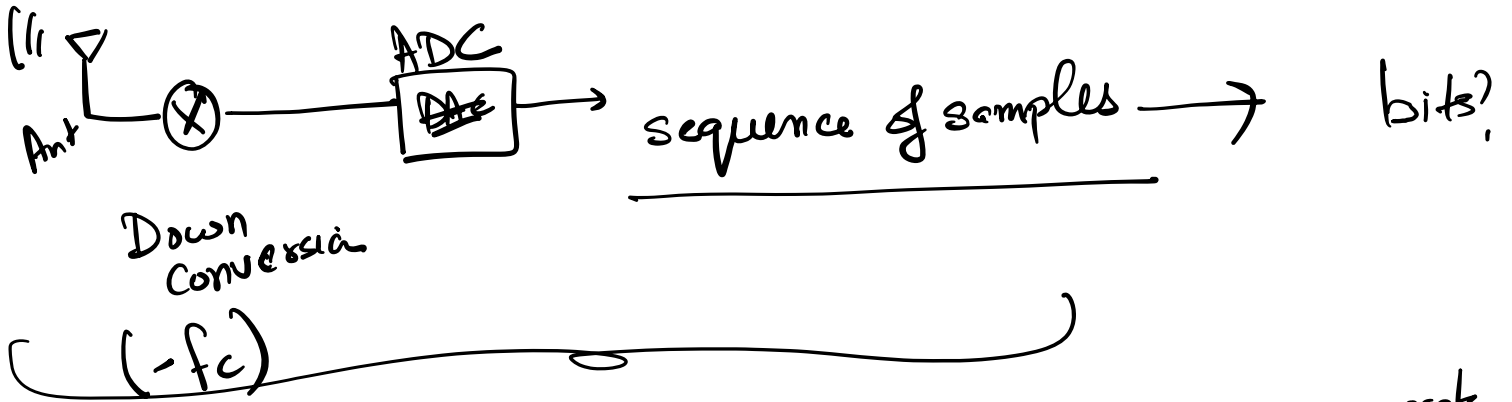
$\omega = 2\pi f$

$\frac{20\text{MHz}}{64} = \frac{20\text{MHz}}{N} \rightarrow 64$  samples

Channel in freq. domain

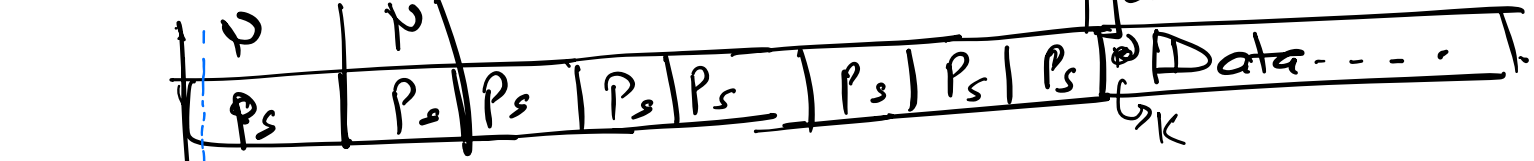


# Packet Detection



Wi-Fi, 64 narrowband freq  $\rightarrow$  1 OFDM symbol  
 64 samples  $\rightarrow$  1 OFDM symbol

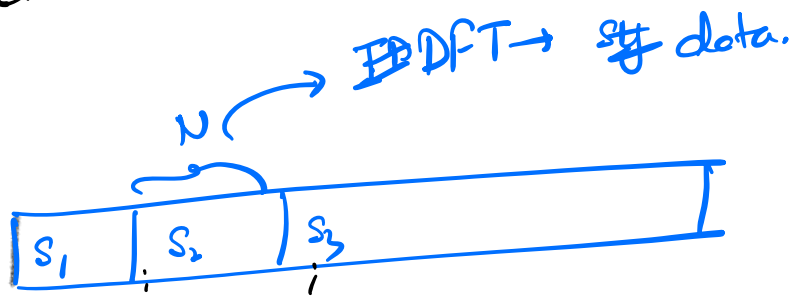
DEP



2 symbol lengths  
 start of the packet

2 samples

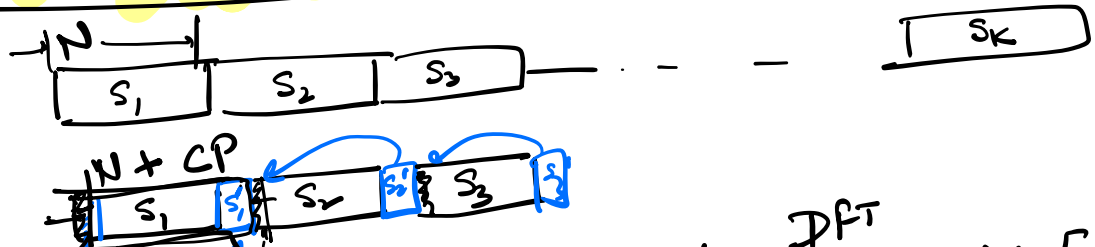
DFT → demodulated  
 (channel) ↓  
 bits.



DFT → X wrong data points

# Cyclic Prefix

CP = length of cyclic prefix.



error tolerance of CP symbols

$$x(t) \xrightarrow{\text{DFT}} X[f]$$

$$x(t - \delta) \rightarrow X[f] e^{-\frac{2\pi f \delta}{N}}$$

Wi-Fi → N = 64  
20MHz

CP = 16  
↳ 25%

$$\frac{64}{64+16} = \frac{64}{80} \times 100\% = 80\%$$

↳ 20%

LTE → N = 1024  
CP = 72

overhead = 6.5%

$$\frac{72}{72+1024}$$

good CP  
↳ error tolerance  
↳ ISI → inter symbol interference.

bad CP  
↳ overhead

0.5 CP into the packet  
or 0.75 CP





# Carrier Frequency Offset

Intuition  
Correction

Upconversion at sender

$$[ + f_c ]$$

$$x(t) e^{+j 2\pi f_c t}$$

↓  
clocks

downconversion

$$[ - f'_c ]$$

~~$$x(t)$$~~  

$$y(t) e^{-j 2\pi f'_c t}$$

↓  
clocks

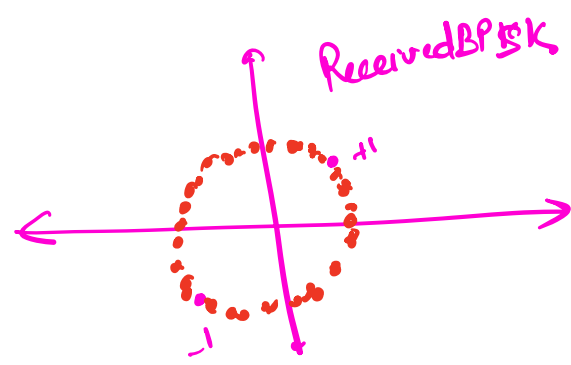
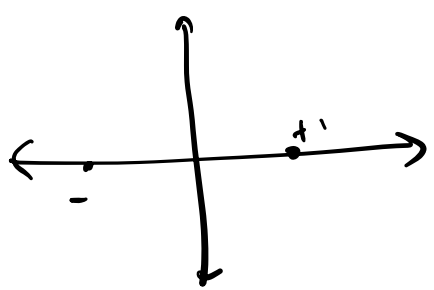
$$y(t) = x(t) e^{+j 2\pi f_c t} e^{-j 2\pi f'_c t}$$

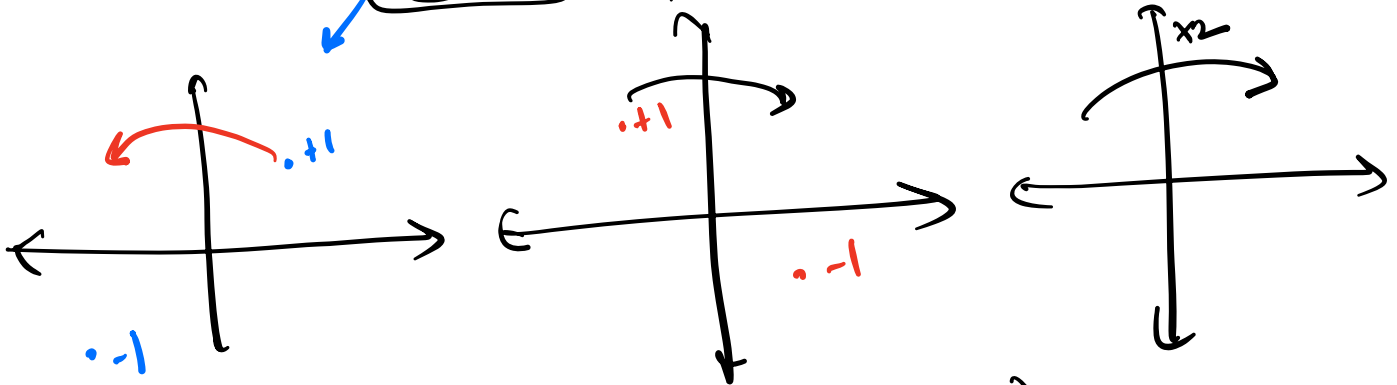
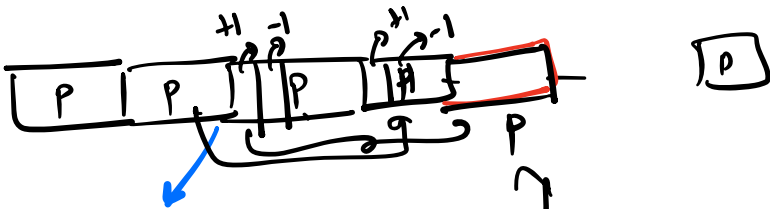
$$= x(t) e^{j 2\pi (f_c - f'_c) t}$$

↓

CFO or carrier freq. offset

BPSK





$$x(t) e^{j 2\pi \frac{(f_c - f_c')}{\Delta f} t}$$

$$y_1(t) = x(t) e^{+j 2\pi \Delta f t}$$

$$y_2(t) = x(t) e^{+j 2\pi \Delta f (t+N)}$$

$$A = \sum_{t=1}^N y_1(t) y_2^*(t) = \sum |x(t)|^2 e^{-j 2\pi \Delta f N}$$

$$\angle A = -j 2\pi (\Delta f) N$$

$$y(t) * e^{-j 2\pi \Delta f N}$$

$$A: y_1(t) = x(t) e^{j2\pi \Delta f t}$$

$$y_2(t) = x(t) e^{j2\pi \Delta f (t+N)}$$

$$\frac{y_2(t)}{y_1(t)} = \underline{\underline{e^{j2\pi \Delta f N}}}$$

$$\frac{y_2(t)}{y_1(t)} = 1$$

$$\underline{\underline{2\pi \Delta f N}} = \underline{\underline{\phi}}$$

$$y_1(t) \quad y_2^* \xrightarrow{\text{negative}} (t) = \underline{\underline{x(t) x^*(t)}}$$

$$|x(t)|^2 e^{j2\pi \Delta f t - 2\pi \Delta f (t+N)}$$

$$\underline{\underline{|x(t)|^2 e^{-j2\pi \Delta f N}}}$$

# Channel Estimation

Residual CFO/SFO



# Summary

## At TX:

- Create preamble symbol from training sequence (Uses BPSK)
- Repeat preamble symbol:
  - 4 times for packet detection
  - 2 times for CFO estimation
  - 2 times for channel estimation
  - Add CP for the last preamble
- Create data symbol from:
  - Data bits (Uses BPSK, QPSK, M-QAM)
  - Pilot bits (Uses BPSK)
- Add cyclic prefix to data symbols.



## At RX:

- Detect beginning of packet.
- Estimate & correct for CFO.
- Jump  $\approx 0.75 CP$  samples into symbol to avoid ISI
- Estimate the channel.
- For each subsequent data symbol:
  - Remove CP
  - Take FFT of Size N
  - Correct for channel by dividing with  $\tilde{H}(f)$
  - Use linear regression to estimate residual CFO and SFO
  - Estimate accumulated phase  $\Delta\phi(f)$  for each frequency bin
  - Add  $\Delta\phi(f)$  to channel estimate  $\tilde{H}(f)$
  - Decode Bits

